



# Achieving Electrothermal Stability in Interconnect Metal During ESD Pulses

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# Outline

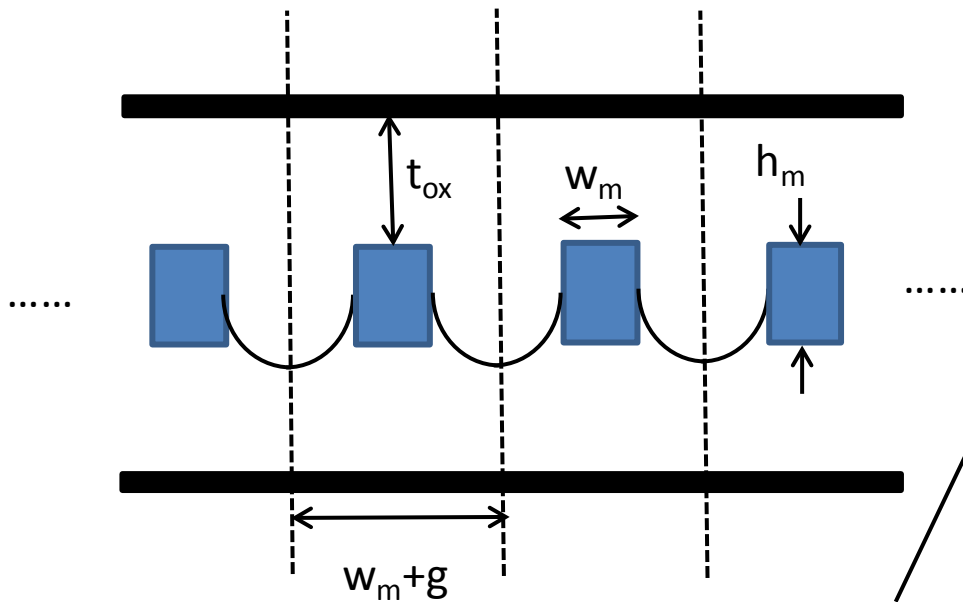
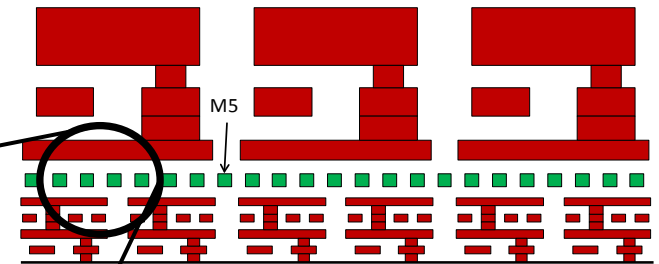
- Partitioned wires as a heating slab
  - More surface/volume than ever
  - Neighboring metal is a heat sink
  - 1-dimensional R-C transmission line models for thermal behavior
- Feedback model for metal heating
  - Self-consistent  $T(t)$  expression
    - From metal tempco and thermal Ohm's Law;  $T(s)=P(s)Z(s)$
    - Positive (tempco) and negative (electrical circuit, e.g., TLP) feedback
- Thermal impulse response  $Z(s) \leftrightarrow Z(t)$ 
  - **Pre-silicon:** finite element modeling
    - Also thermal circuit models and known limiting conditions
  - **Post-silicon:** Transmission Line Pulse (TLP) measurements
    - Derive  $Z(s)$  from  $T(s)=P(s)Z(s)$ , having measured  $T(t)$  and  $P(t)$ 
      - Pole-zero expression for complex thermal impedance
        - Corresponds to 5-element R-C model
- ESD predictions in Excel using  $Z(t)$  and convolution software
  - Convolve  $Z(t)*P(t)$  for HBM and CDM ESD conditions
    - Obtain complete  $T(t)$  waveform from feedback equation and check for melting



# Wires Embedded in ILD Oxide

Thermal Ohm's Law:

$$T(s) = P(s)Z(s)$$

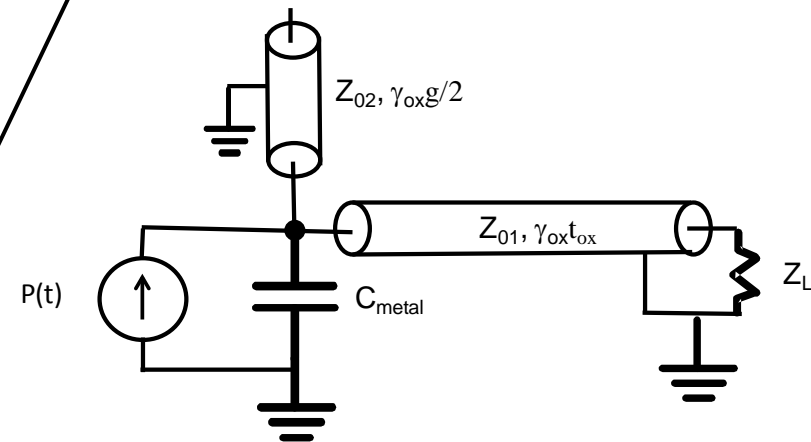


Pattern C as shown

Pattern D is wider,  $g_D = 2g_C$

Note open circuit b.c.

Thermal circuit model:



# Thermal Feedback Model

metal resistance  $R(t) = R_0(1 + \alpha T(t))$   $\alpha = \text{metal tempco} = 0.0025/^\circ\text{C}$  for Cu

$$T(t) = P_0(t) * Z(t) + \alpha [T(t)P_0(t)] * Z(t) \quad \text{thermal Ohm's Law}$$

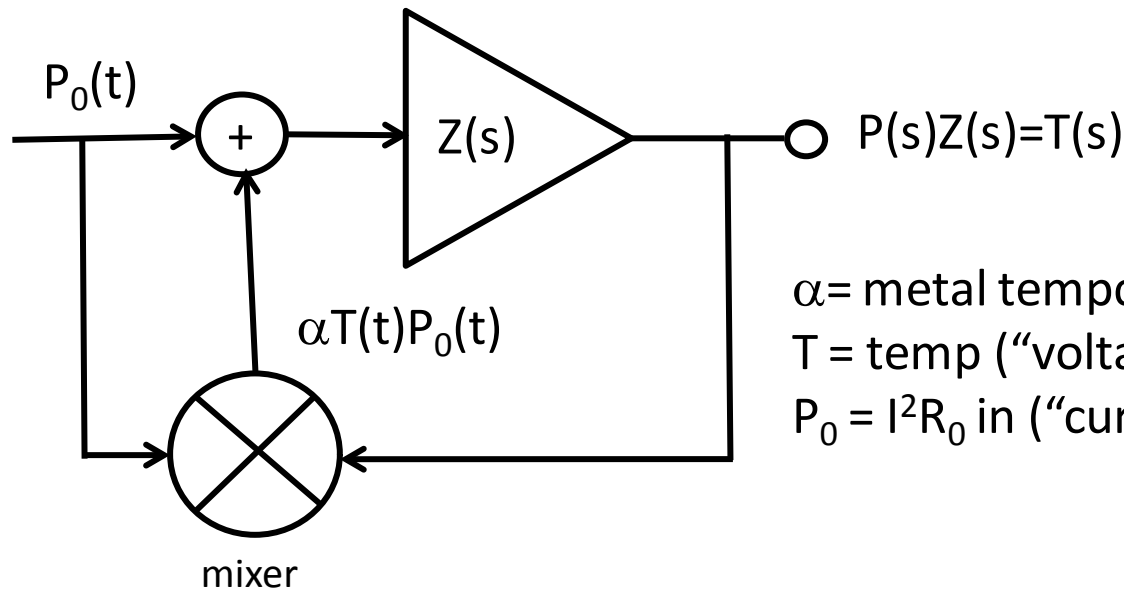
or

$$T(t) = \frac{P_0(t) * Z(t)}{1 - \alpha \frac{[T(t)P_0(t)] * Z(t)}{T(t)}}$$

$$P_0(t) * Z(t) = \int_0^t P_0(t-\tau)Z(\tau)d\tau \Leftrightarrow P_0(s)Z(s)$$

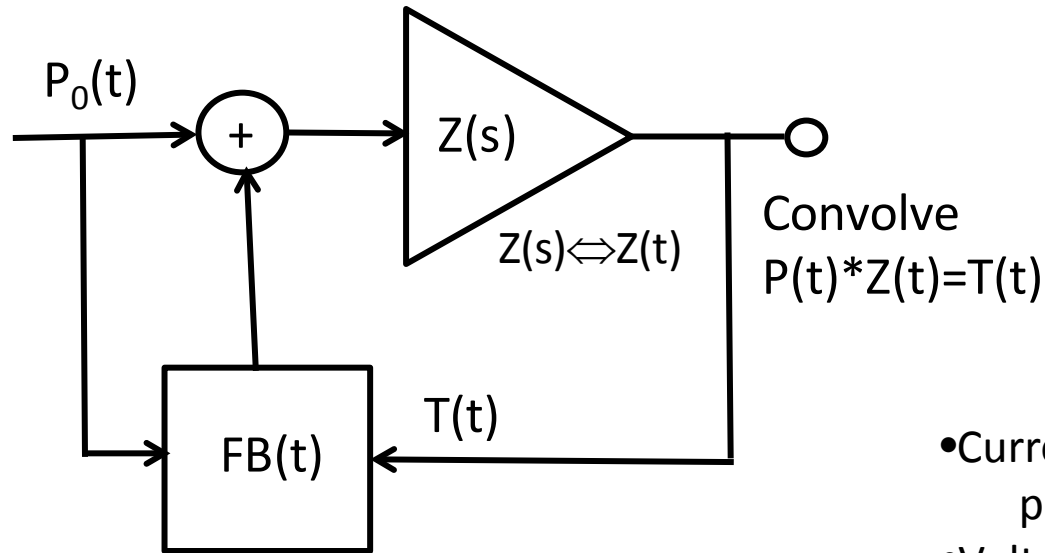
Solve implicit equation for T(t)

For  
"current"  
source  
 $P_0(t)$ :



$\alpha = \text{metal tempco}$   
 $T = \text{temp ("voltage")}$   
 $P_0 = I^2 R_0$  in ("current")

# General Feedback Network



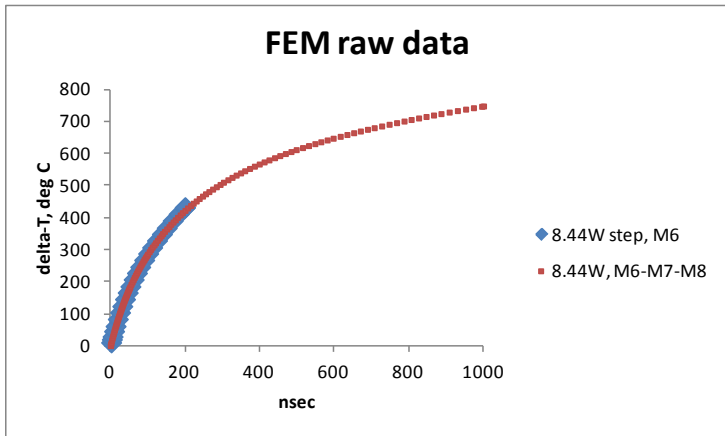
- Current source produces positive feedback
- Voltage source produces negative feedback

For TLP:

$$P_0(t) = P_0 = \frac{V_0^2 R_0}{(R_0 + 50)^2} \quad FB(t) = P_0 \left[ \left( \frac{1 + \alpha T(t)}{\left( 1 + \frac{\alpha R_0 T(t)}{R_0 + 50} \right)^2} \right) - 1 \right]$$

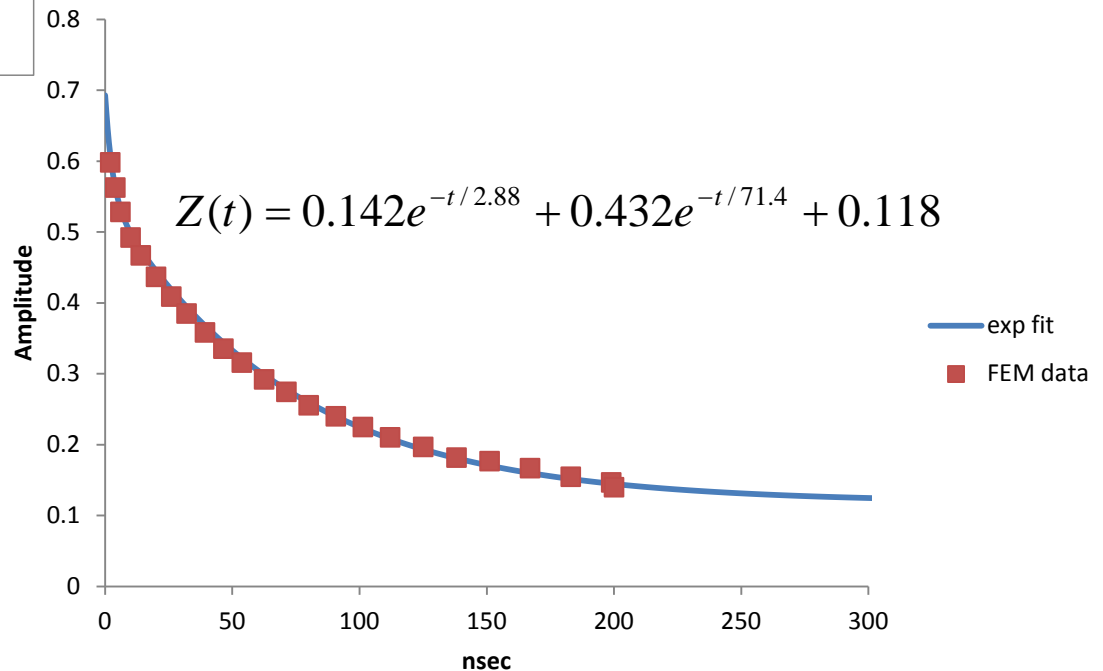
- Because of source resistance  $Z_s=50\Omega$ , TLP introduces negative feedback, and when  $R(t)>50\Omega$ , it becomes net negative

# Z(t) from Finite Element Modeling

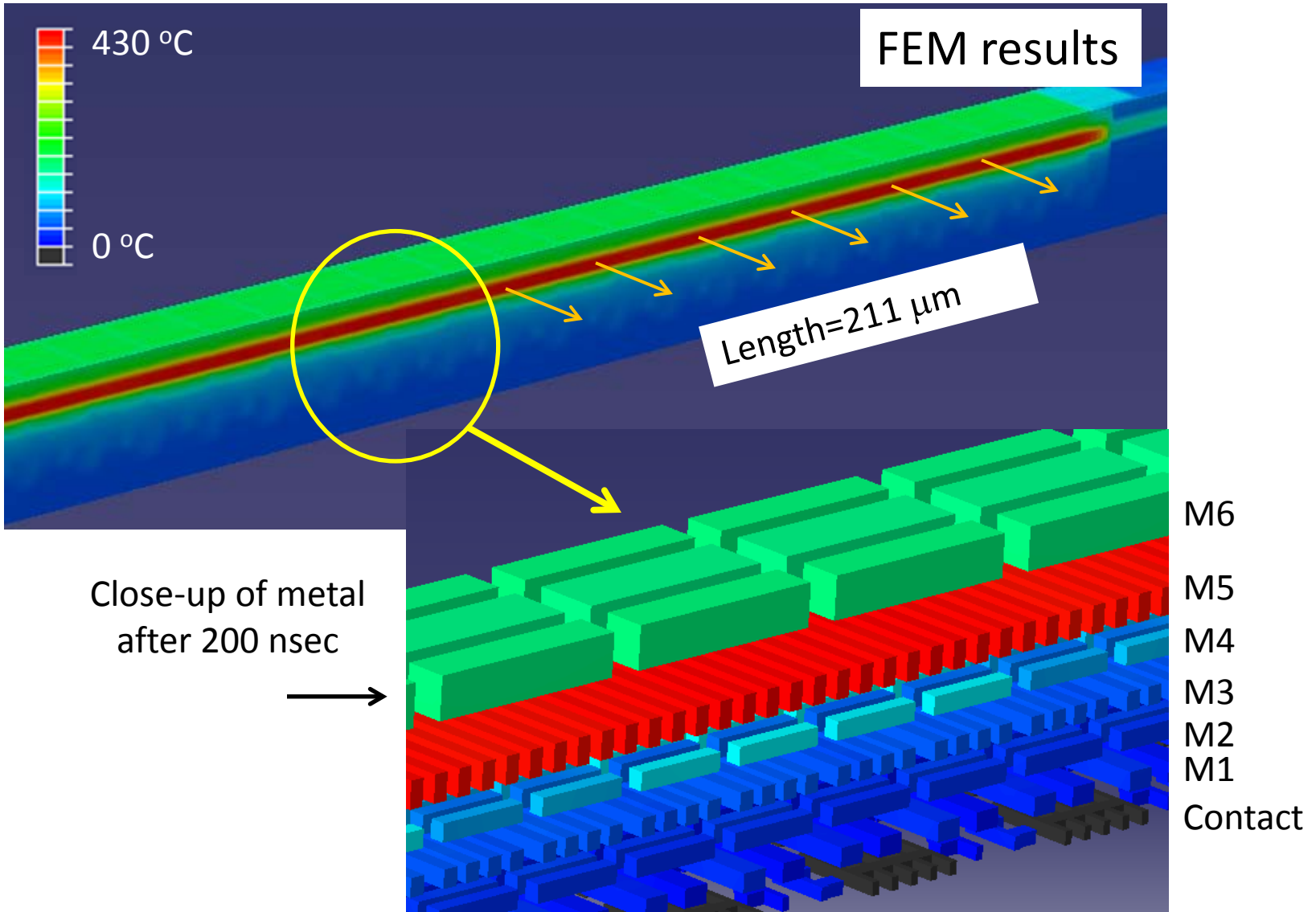


Step response gives **thermal impulse response Z(t)**

differentiate  
and normalize



Final offset (0.118) substitutes for LONG (250 nsec, 1  $\mu$ sec) time constant terms





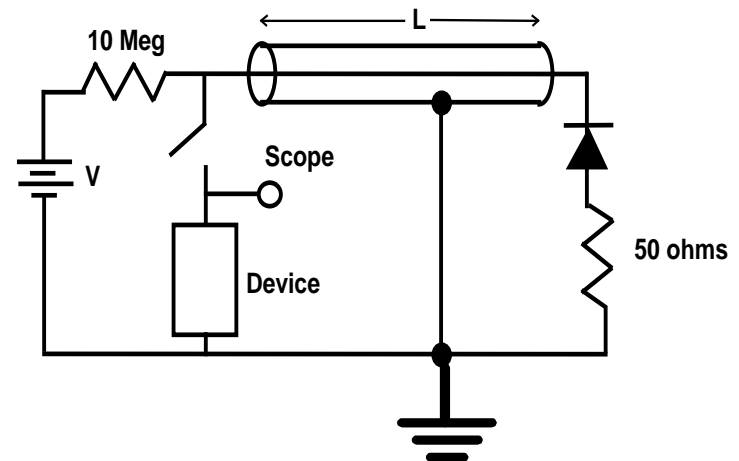
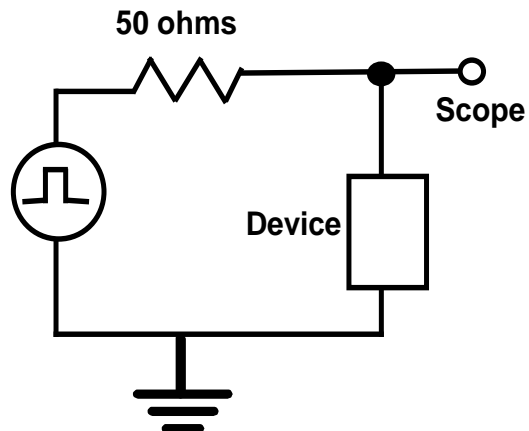
# Transmission Line Pulsing (TLP)

- Transmission line pulsing generates brief, high current (several ampere) pulses; same current/time scale as ESD

Setup:

- Equivalent circuit

( $R_{\text{device}} < 50 \Omega$ ):

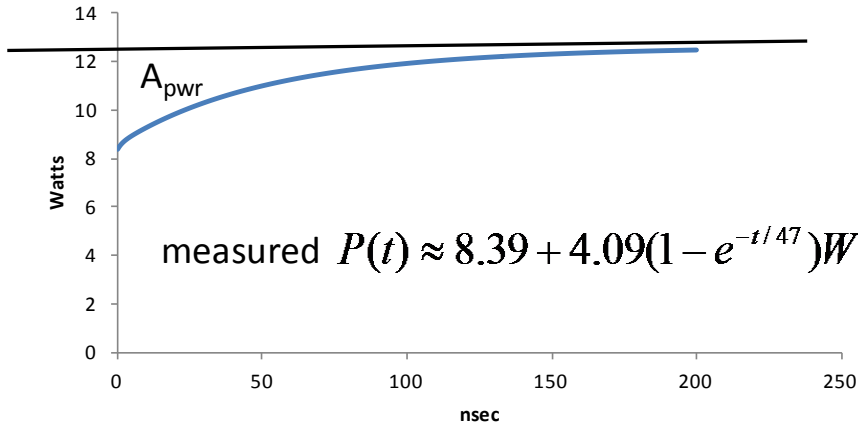


$$I_{\text{device}} = (V - V_{\text{device}}) / 50$$
$$t_{\text{pulse}} = 2L/c, \quad c = 20 \text{ cm/nsec}$$

Apply to metal lines and use Cu tempco ( $\alpha = 0.0025$ ) for *in-situ* T measurement

# TLP data

TLP Power, M5-C-60V



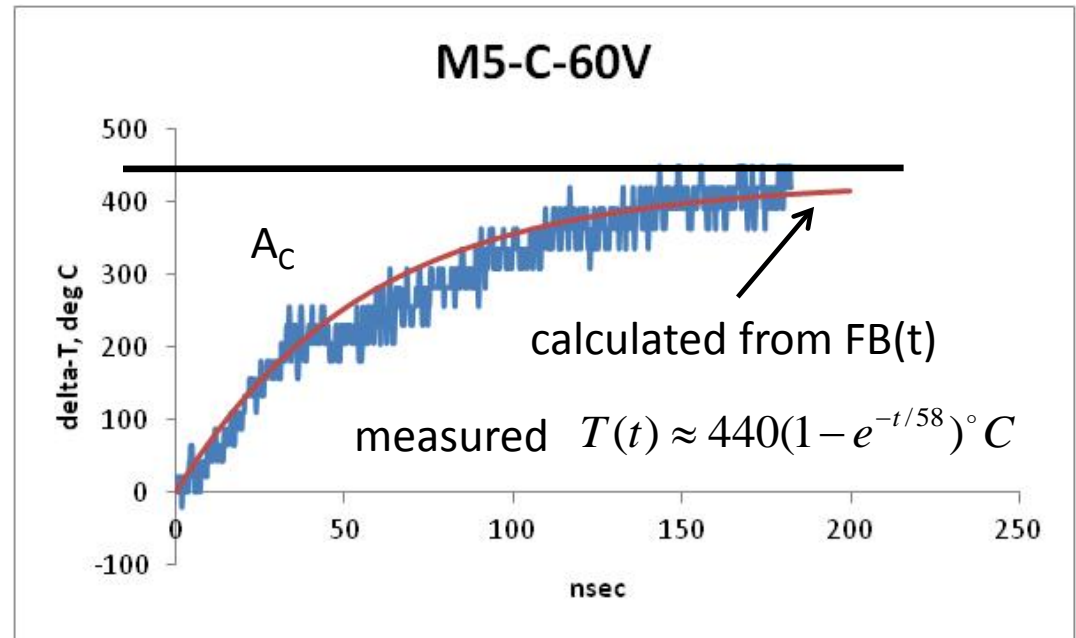
Pattern C, 60V

$$P(s) = \frac{8.39W}{s} + \frac{4.09W}{s(1 + 47s)}$$

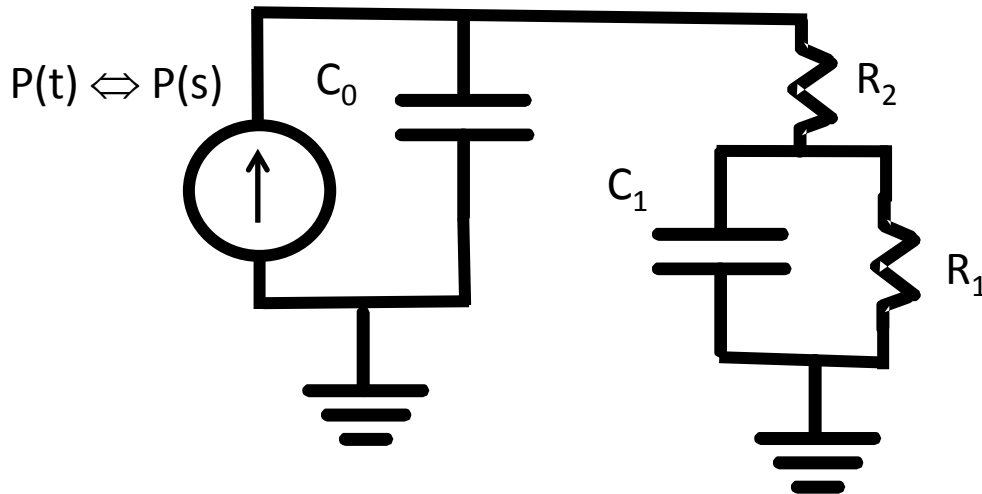
normalized  $A_c$ ,  $A_{pwr}$  give  
time constants

$$T(s) = \frac{440^\circ C}{s(1 + 58s)}$$

$$Z(s) = \frac{T(s)}{P(s)}$$



# Thermal Impedance



$$C_0 = 1.1 \text{ nJ/}^\circ\text{C (metal + oxide)}$$

$$R_2 = 32.7 \text{ }^\circ\text{C/W (oxide)}$$

$$C_1 = 20 \text{ nJ/}^\circ\text{C}$$

$$R_1 = 2.52 \text{ }^\circ\text{C/W}$$

Values for Pattern C, 60V  
from slide 10

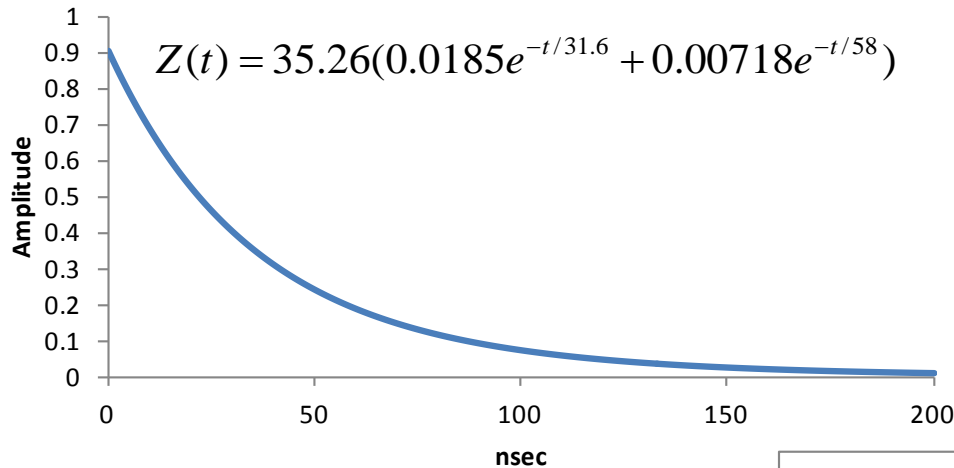
$$Z(s) = \frac{(R_1 + R_2)(1 + R_1 \parallel R_2 C_1 s)}{1 + (R_1 C_1 + (R_1 + R_2) C_0) s + R_1 R_2 C_1 C_0 s^2}$$

$$P_0 = \frac{V_0^2 R_0}{(50 + R_0)^2} \text{ for TLP}$$

- $Z(s)$  from slide 10 has 2 poles and 1 zero; 5-element RC network
- For TLP, temp should flatten out at  $P_{\text{final}}(R_1 + R_2) = P_{\text{final}} Z_0 = T_{\text{final}}$

# Thermal Impulse Response

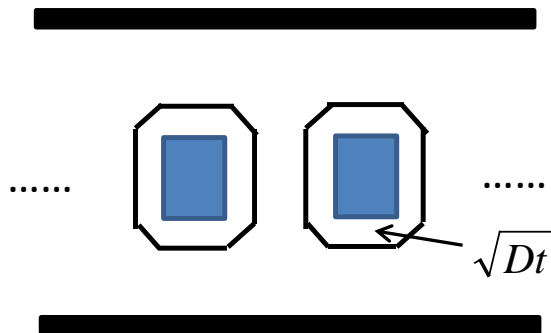
## Z(t) from TLP data (°C/nJ)



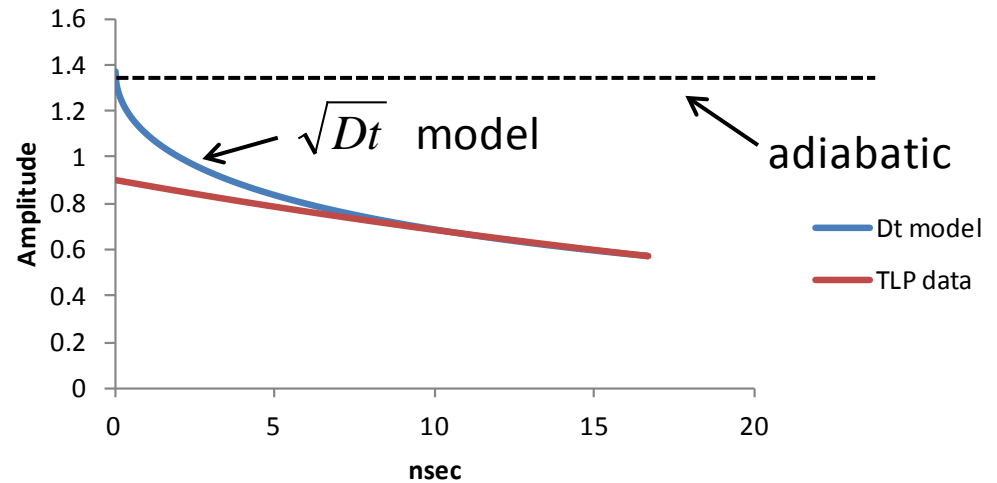
from  $Z(s) = 35.26 \frac{1+47s}{(1+31.6s)(1+58s)} \text{ } ^\circ\text{C/W}$

$$= \frac{T(s)}{P(s)}$$

thermal sheath for short time:

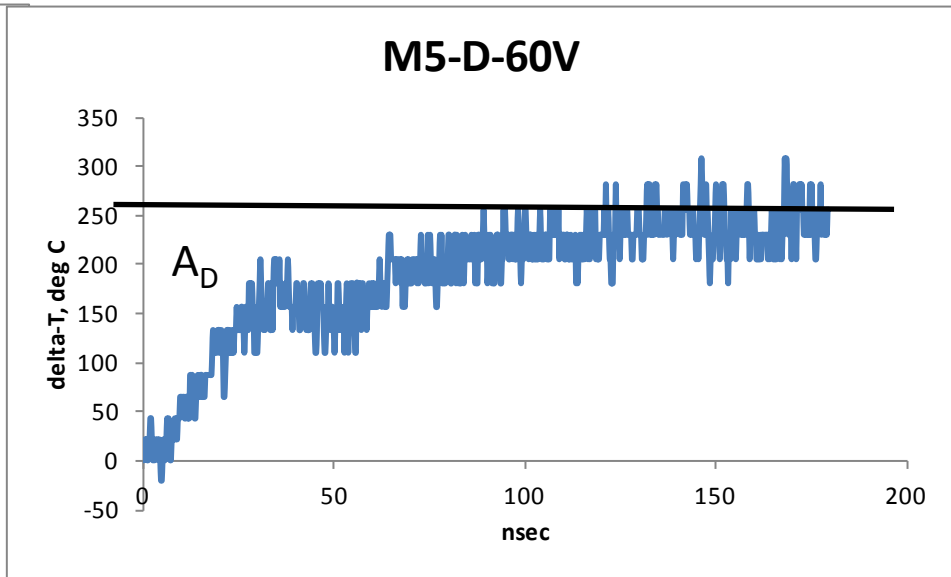
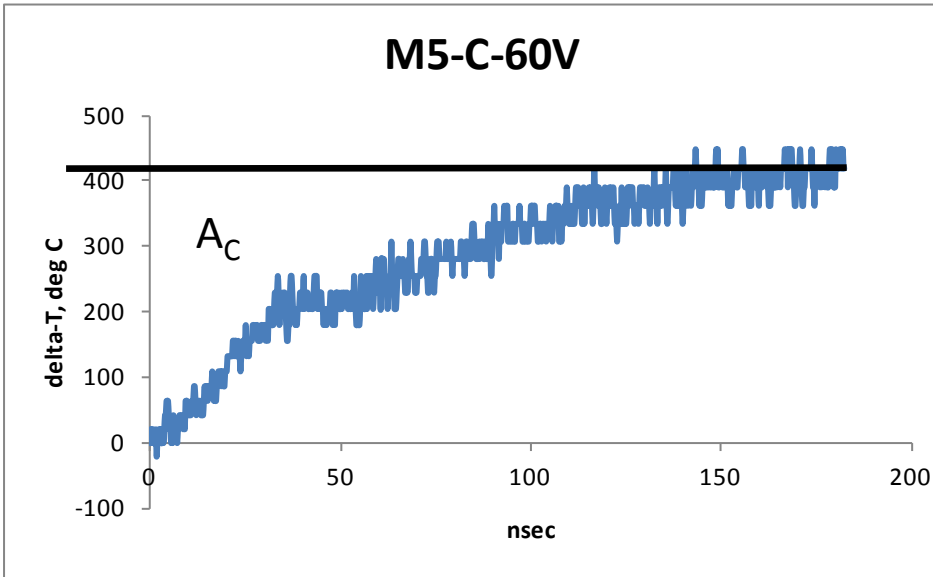
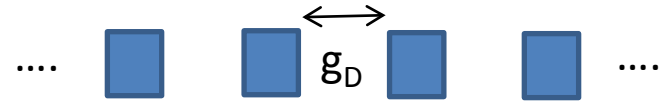
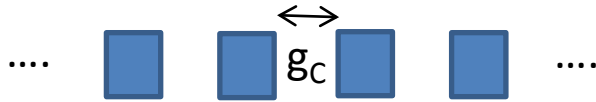


## Z(t) at short times



adiabatic limit is from Cu line heat capacity;  $1.37 \text{ } ^\circ\text{C/nJ}$

# TLP on Patterns C and D, 60V, 1 $\mu\text{m}^2$ cross-section



Pattern C width =  $X$   
 $T_{final} = 440$  °C (“volts”)  
 $Z_{OC} = 35.26$  °C/W (“ohms”)  
 $P_0 = 8.24\text{W}$ ,  $P_{final} = 12.16\text{W}$   
 $A_C$  (normalized) = 58 nsec

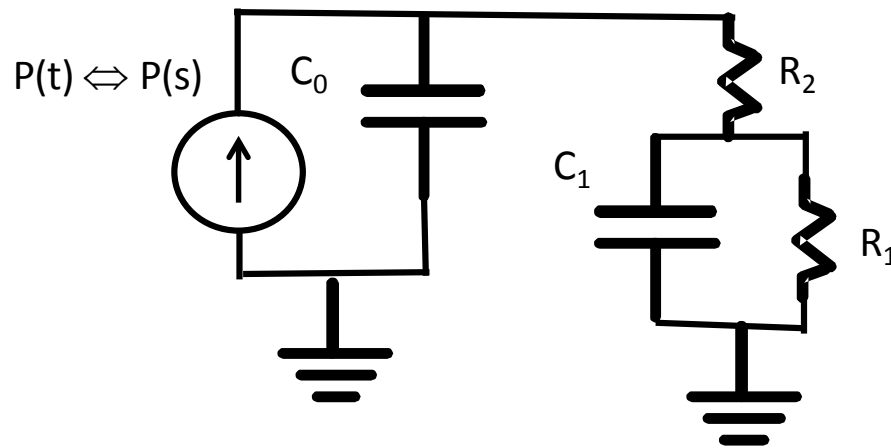
$\alpha = 0.0025$

Pattern D width =  $1.48X$  ( $g_D = 2g_C$ )  
 $T_{final} = 225$  °C (“volts”)  
 $Z_{OD} = 23.2$  °C/W (“ohms”)  
 $P_0 = 8.24\text{W}$ ,  $P_{final} = 10.99\text{W}$   
 $A_D$  (normalized) = 46 nsec

# Thermal Circuit Elements

Pattern	TLP Volts	$C_0$	$R_2$	$R_1$	$C_1$	$T_{final}$
C	60	1.1 nF	32.7 $\Omega$	2.52 $\Omega$	20 nF	440 $^{\circ}\text{C}$
C	70	1.09 nF	33.9 $\Omega$	2.84 $\Omega$	21.7 nF	645 $^{\circ}\text{C}$
D	60	1.53 nF	21.9 $\Omega$	1.33 $\Omega$	29.5 nF	225 $^{\circ}\text{C}$

from TLP data



$$\Omega \Rightarrow ^{\circ}\text{C}/\text{W}$$

$$\text{nF} \Rightarrow \text{nJ}/^{\circ}\text{C}$$

Trend for wider Pattern D  
( $g_D = 2g_C$ ) is as expected

# Stability Criterion, Step Current

Eq. (10), steady state:

$$T_{final} = P_{final} Z_{0th} = \frac{P_0 Z_{0th}}{1 - \alpha P_0 Z_{0th}}$$

$$J = \frac{i}{\ell W} \quad R_{el} = \rho_0 \frac{z}{\ell W}$$

$$P_0 = i^2 R_{el} = \frac{J^2 \rho_0 z \ell^2 W^2}{\ell W} = J^2 \rho_0 z \ell W$$

electrical

$$Z_{0th} = \frac{1}{K_{ox}} \frac{\delta}{z W}$$

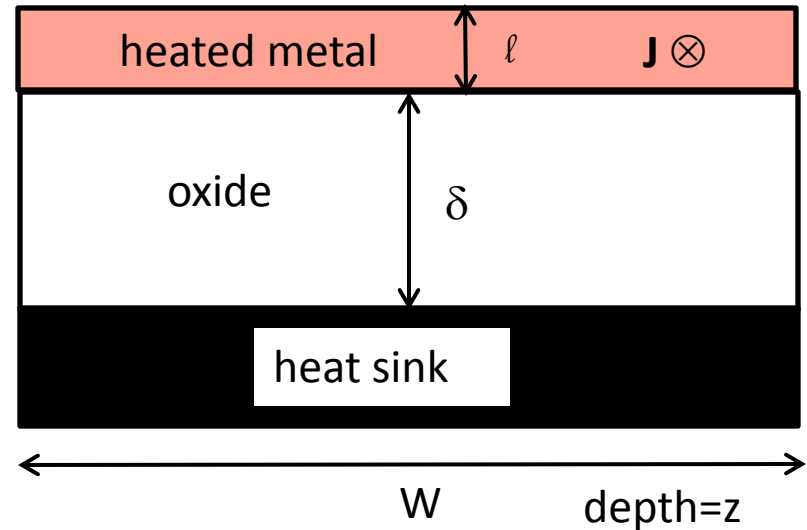
thermal

$T_{final}$  is finite unless

$$\alpha P_0 Z_{0th} > 1$$

or

$$\frac{\alpha J^2 \rho_0 \delta \ell}{K_{ox}} > 1$$



D.G. Pierce (1982) derived the same condition for “unbounded” solutions:

$$J^2 > \frac{K_{ox}}{\alpha \rho_0 \delta \ell}$$

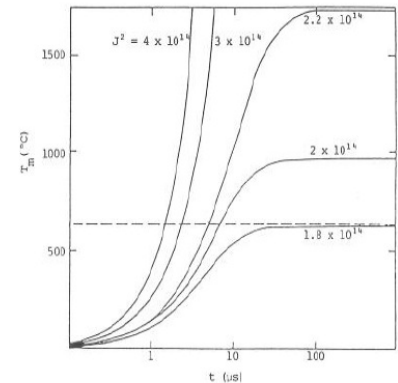
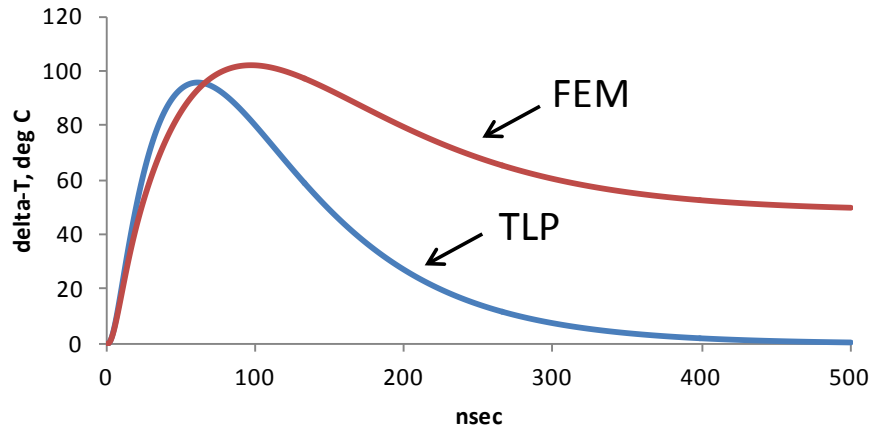


Figure 4. Temperature Versus Time for Aluminum Metallization

# HBM Temperature Waveforms

**HBM 1kV, Pattern C**



Convolve  $Z(t)$  with Human Body Model  $P_0(t)$  function and solve feedback equation for self-consistent  $T(t)$  in Excel

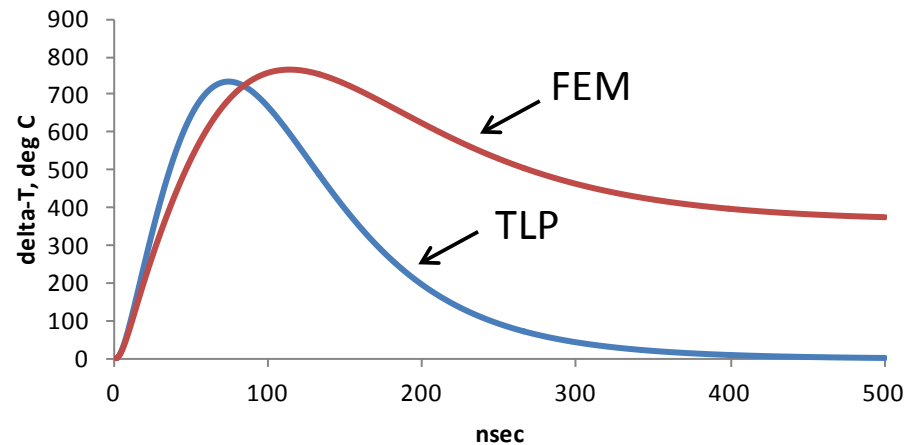
$T_{\max}$  is similar despite the ultimately “capacitive” FEM  $Z(t)$

For 1kV HBM,

$$P_0(t) = \frac{1040}{201} \left[ -2e^{-t/4} + e^{\frac{\sqrt{201}-15}{60}t} + e^{\frac{\sqrt{201}+15}{60}t} \right]$$

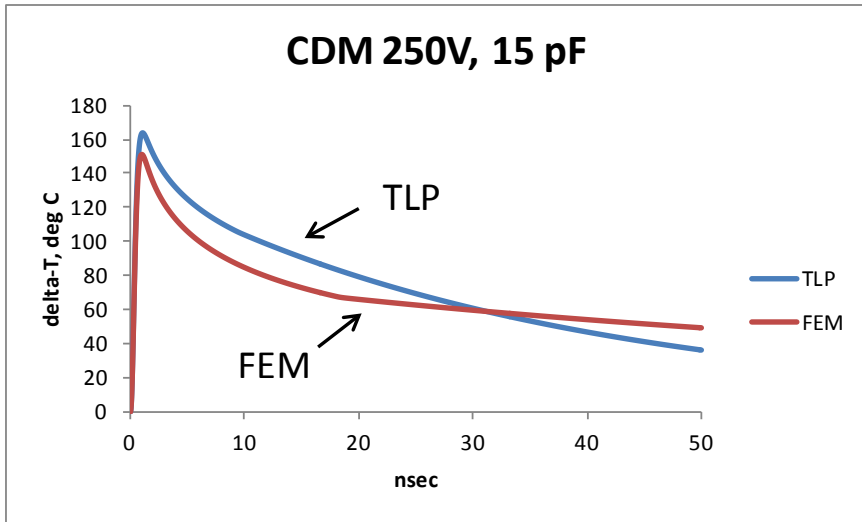
$P_{0-2kV}(t) = 4P_{0-1kV}(t)$  but  $T_{\max}$  is higher due to positive feedback effect

**HBM 2kV, Pattern C**





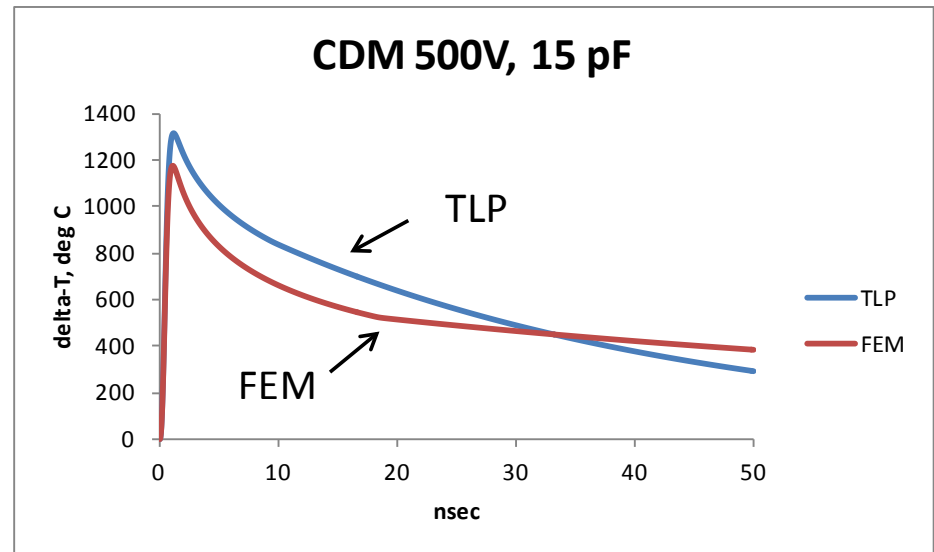
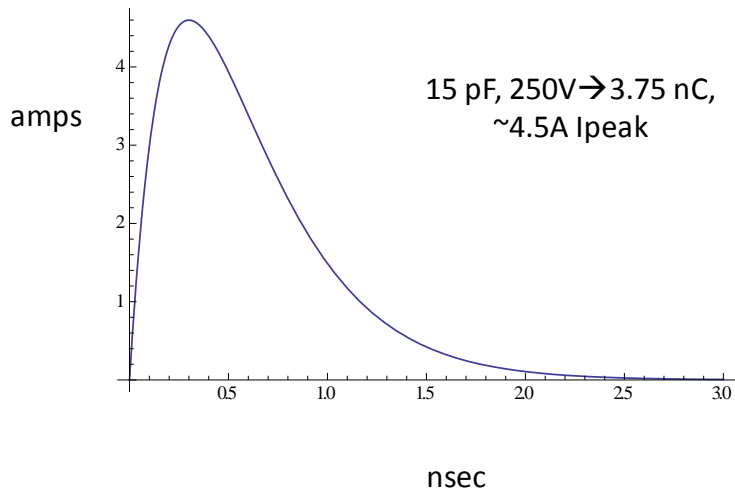
# CDM Temperature Waveforms



Using  $Z(t)$  for Pattern C,  $1 \mu\text{m}^2$  Cu  
Excel calculations

At 500V ( $I_{\text{peak}}=9\text{A}$ ), feedback pushes Cu  
metal temperature beyond the melting  
point ( $1085 \text{ }^\circ\text{C}$ ) for a few nsec

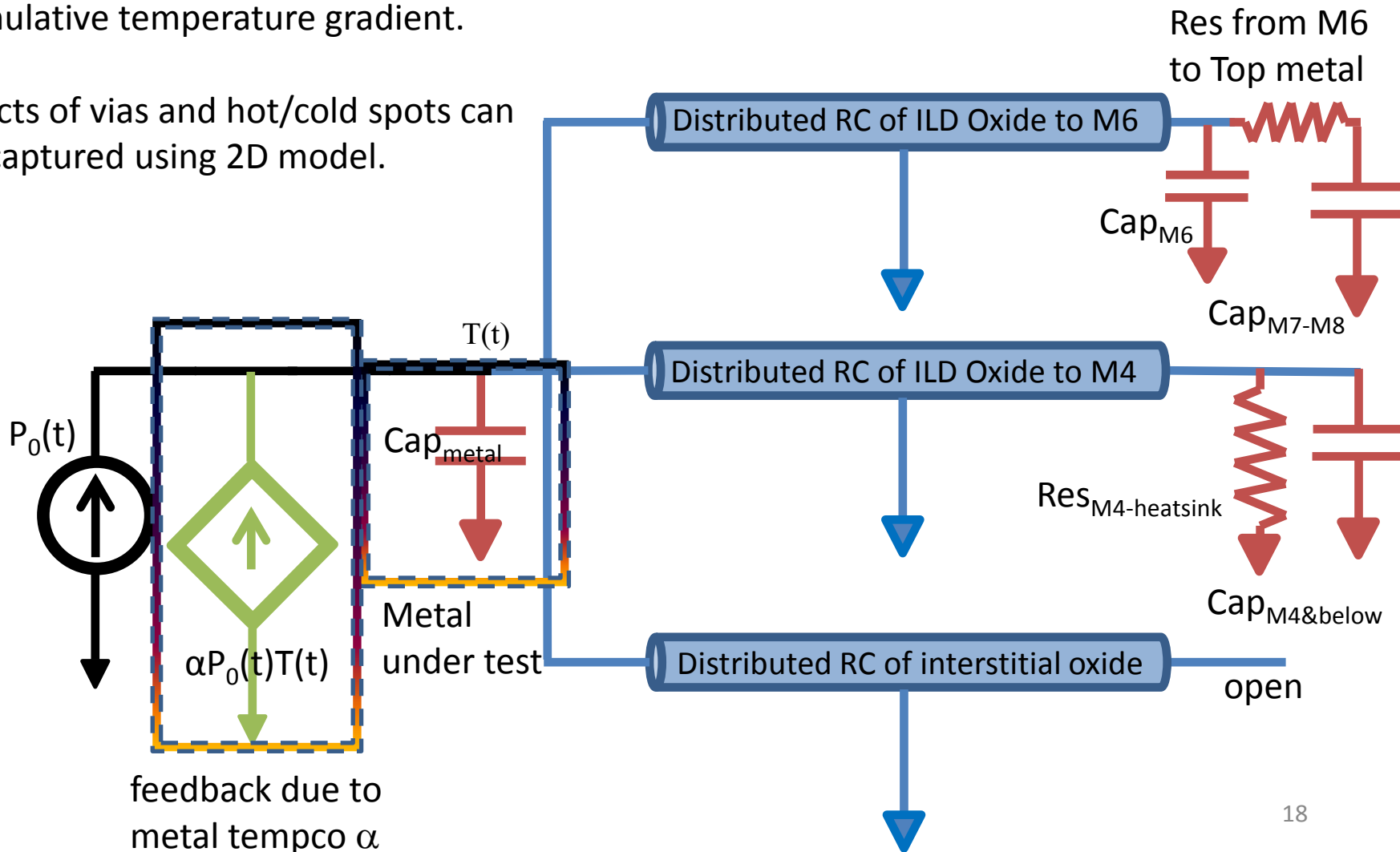
CDM current waveform:



# 1D representation of heat flow

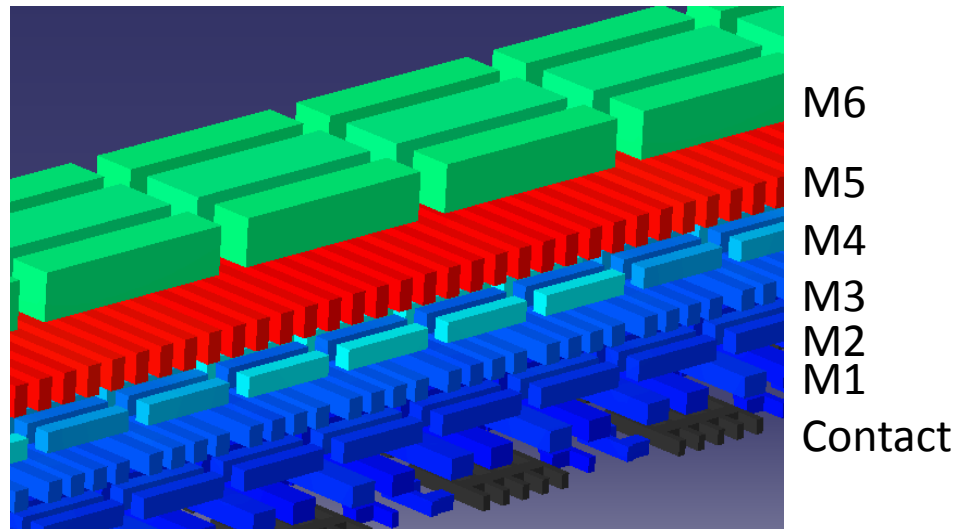
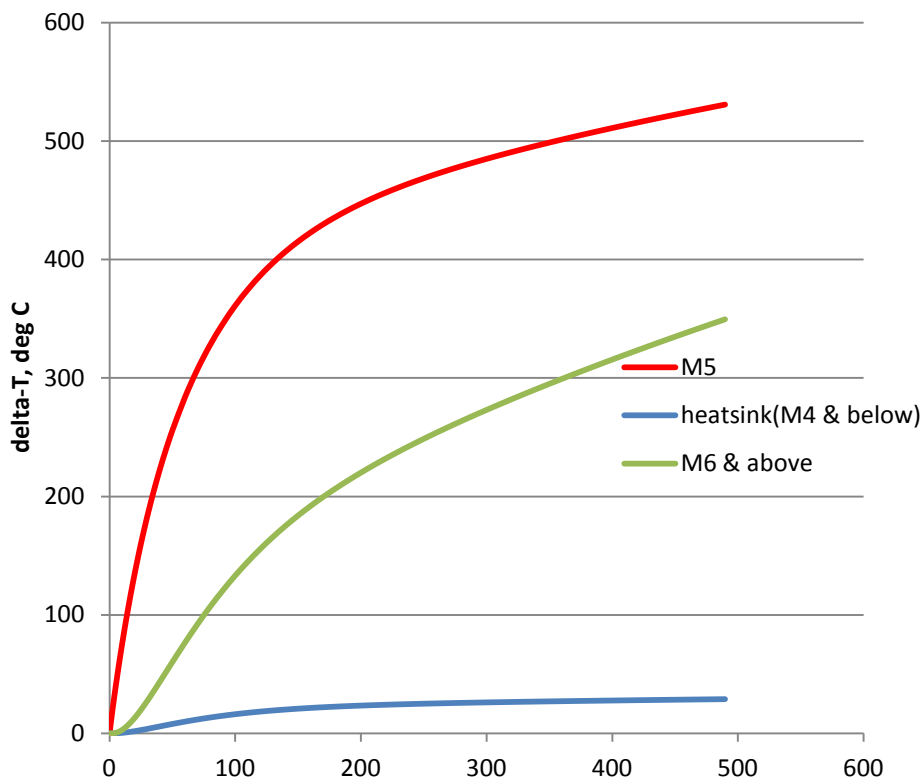
1D **SPICE-like circuit model** captures cumulative temperature gradient.

Effects of vias and hot/cold spots can be captured using 2D model.

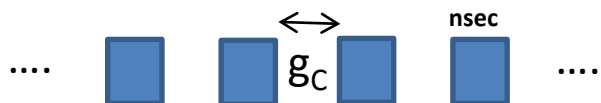


# Results : FEM vs 1D model, step response (pattern C)

M5-C-60V



Close-up of metal after 200 nsec

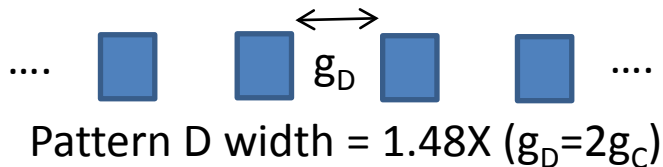
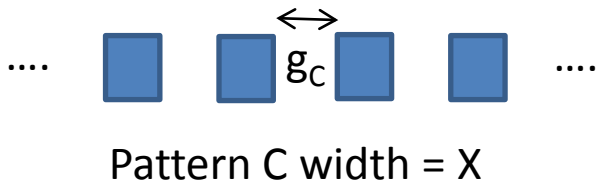


Pattern C width =  $X$   
Power: 8.44W

The temperature profile of M6 & above and M4 & below of the 1D model matches closely with FEM (see slide 7).

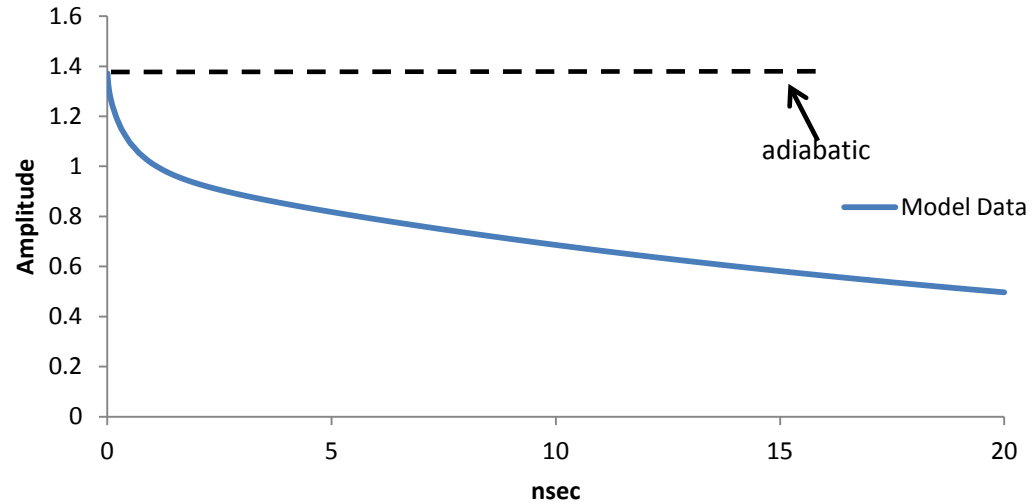
# Impulse Response

From SPICE-like circuit model,  
no feedback

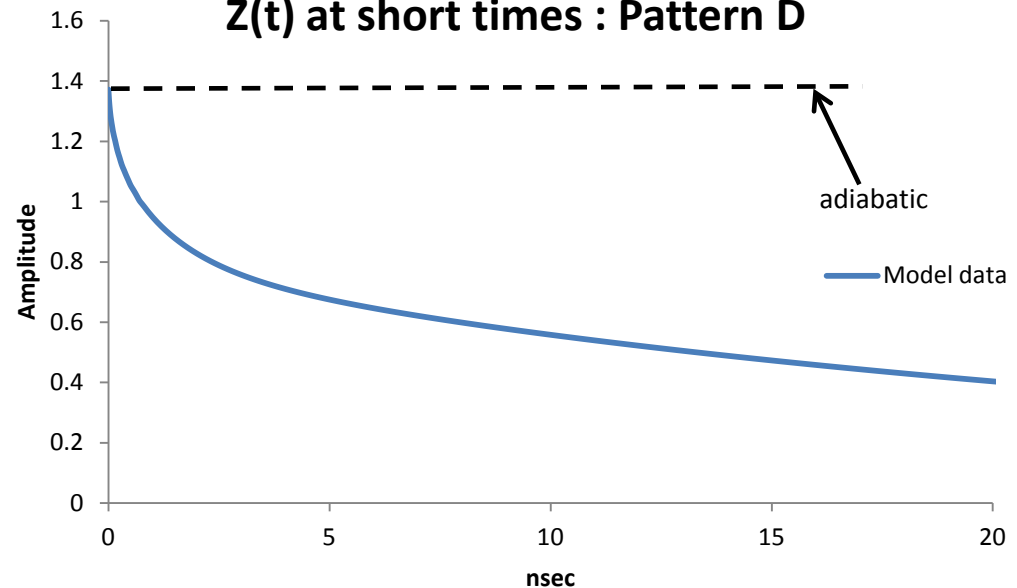


See slide 7 for derivation from step

## Z(t) at short times : Pattern C

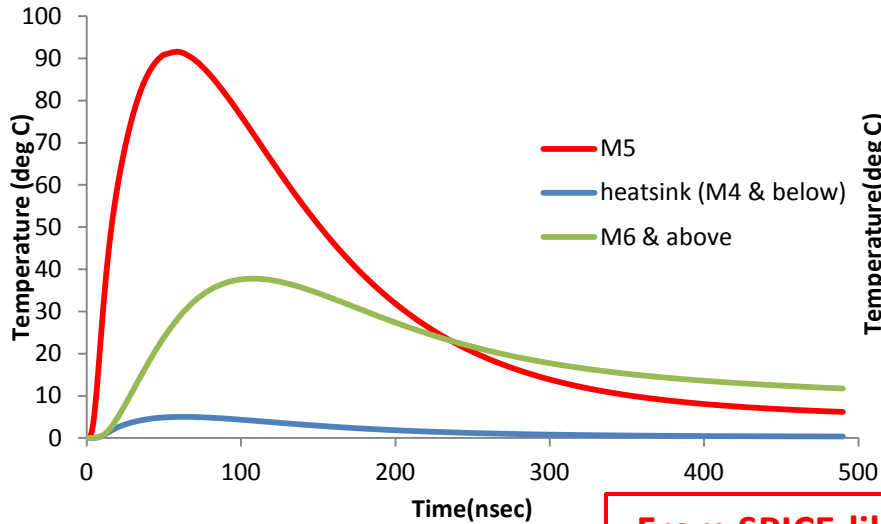


## Z(t) at short times : Pattern D

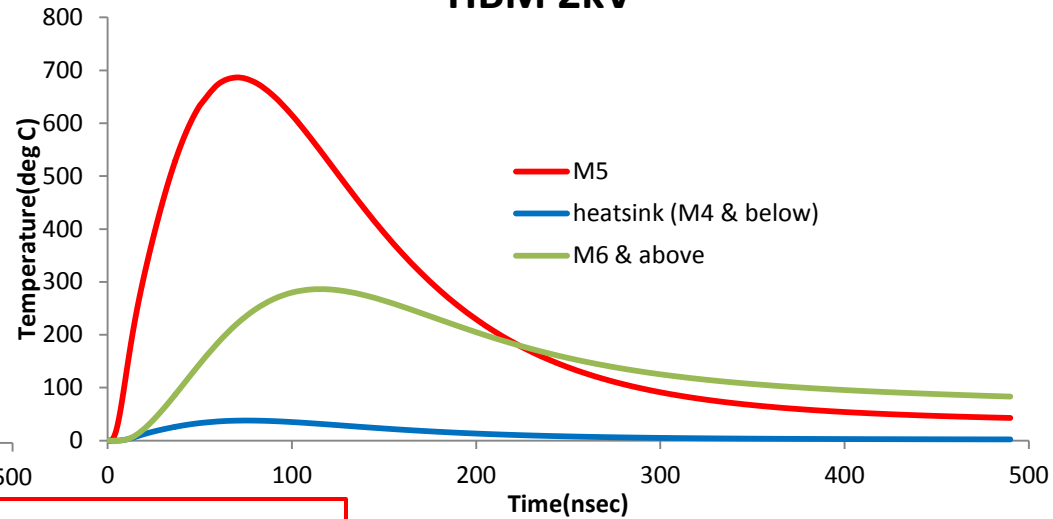


# CDM & HBM response: Pattern C

## HBM 1kV

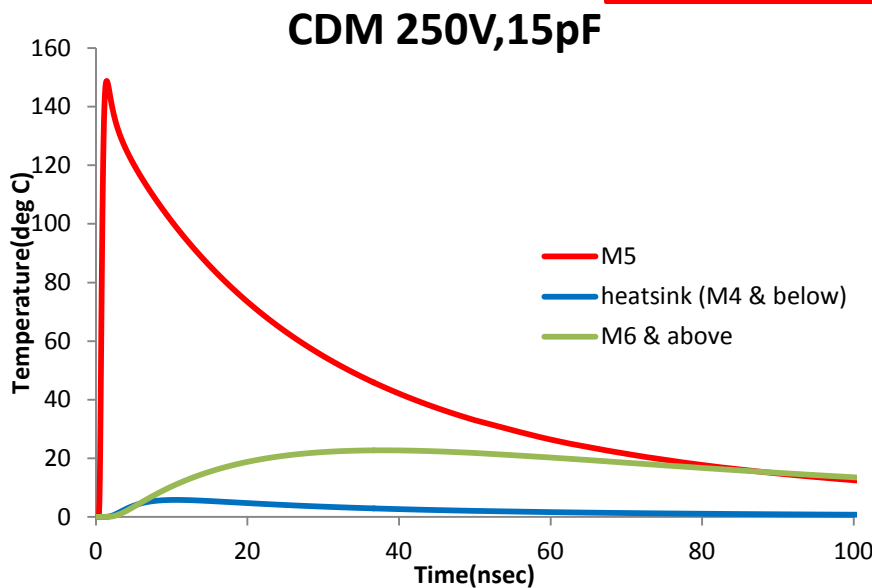


## HBM 2kV

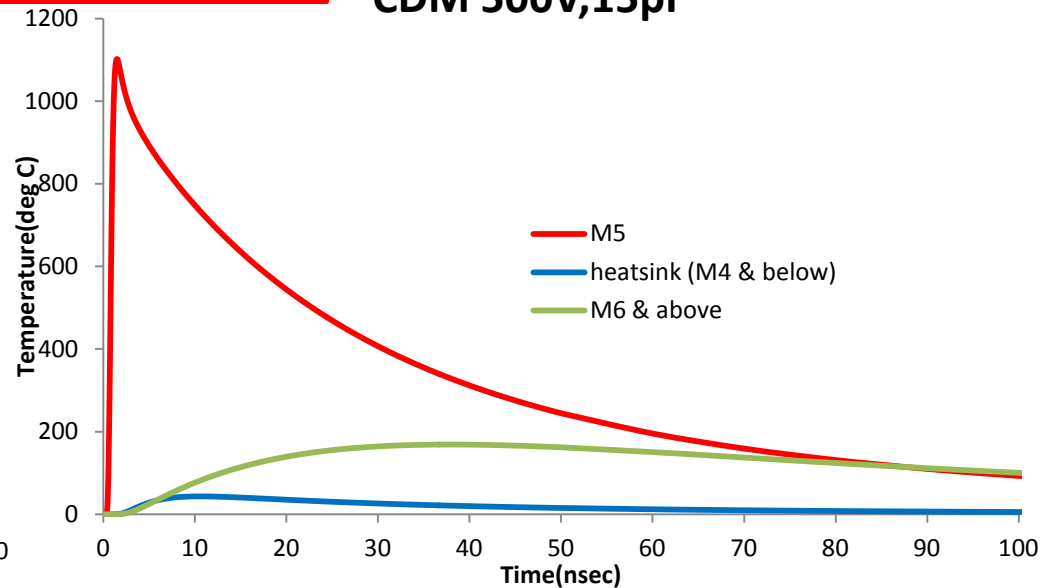


From SPICE-like circuit model,  
with feedback

## CDM 250V,15pF



## CDM 500V,15pF



# Conclusions

- Feedback model for metal heating presented
  - Self-consistent  $T(t)$  expression from thermal Ohm's Law and  $\alpha$ 
    - Positive feedback from pure current source
    - Negative feedback from pure voltage source
      - Electrical source/load impedance determines net feedback
- Thermal impulse response  $Z(s) \leftrightarrow Z(t)$  is central to solving for  $T(t)$ 
  - **Pre-silicon:** finite element modeling (FEM) or SPICE-like circuit model
    - Differentiate response to heat step to get  $Z(t)$
  - **Post-silicon:** Transmission Line Pulse (TLP) measurements
    - Measure  $T(t)$  and input power  $P(t)$ 
      - Maps to 5-element R-C model for each waveform
      - Add simple "heat sheath" model for  $0 < t < 10$  nsec
        - Meshes nicely with TLP data
- ESD predictions in Excel using  $Z(t)$  and convolution software
  - Convolve  $Z(t) * P(t)$ , solve for  $T(t)$  for HBM and CDM ESD conditions
    - $1 \mu\text{m}^2$  Cu x-section gives  $T_{\text{max}} < 800^\circ\text{C}$  for 2kV HBM
    - But 500V CDM (7.5nC) melts Cu with  $T_{\text{max}} \approx 1200\text{-}1300^\circ\text{C}$ 
      - Brief event; should re-solidify but with material changes
    - Strong effect of positive feedback seen for higher ESD voltages
  - SPICE-like circuit model agrees